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IN SOLID CYLINDRICAL RODS OF #1010 STEEL

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N66 28519
(ACCESSION NUMBER)
70
(PAGES)
75741
(NASA CR OR TMX OR AD NUMBER)

(THRU)
1
(CODE)
17
(CATEGORY)

REF ID: A

CFSTI PRICE(S) \$

\$

A paper recommended by the IEEE Rotating Machinery Committee of the IEEE Power Group for presentation at the IEEE Summer Power Meeting, New Orleans, La., July 10-15, 1966. Manuscript submitted April 18, 1966; made available for printing May 19, 1966.

Price: \$1.00
(75¢ at meeting)

(16¢ per copy additional if first class mailing desired)

(26¢ per copy additional if air mail is desired)

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353 East 47th Street
New York, N. Y. 10017

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Paper No.

31 PP 66-434

A NUMERICAL ANALYSIS OF HYSTERESIS AND EDDY CURRENT LOSSES

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INTRODUCTION

Recent studies by the authors of this paper concerning Flux Penetration in Ferromagnetic Material 1,2,3 have led to the development of a computerized procedure for determining the intensity and phase of magnetic field penetration into ferromagnetic materials of various geometries. The present paper extends this prior work to the specific analysis of losses in a cylindrical rod of #1010 steel; the basic procedure and theory, however, are generally applicable to all isotropic ferromagnetic materials. The data used in this procedure are obtained directly from the actual magnetization curve and major hysteresis loop of the material under study. It is believed that the procedure described herein represents a major contribution to the literature of loss analysis.

THEORETICAL BACKGROUND

As shown in reference (2), when a cylindrical rod of radius a is subjected to a known magnetic field intensity axially directed at its surface, $H_z(a,t)$, the finite-difference relationship between magnetic induction, $B_z(r,t)$, and magnetic field intensity, $H_z(r,t)$, as a function of distance (radius) and time, is given by

$$B_z(i,j+k) = \frac{(\Delta t) \rho 10^8}{0.4\pi(\Delta r)^2} \left[H_z(i+h,j) \cdot (1 + \frac{\Delta r}{r}) - 2H_z(i,j) + H_z(i-h,j) \cdot (1 - \frac{\Delta r}{r}) + B_z(i,j) \right], \quad (1)$$

where ρ is the resistivity of the material, k and Δt are the time increments, h and Δr are the space increments, i is the space index, j is the time index, and r is the radius corresponding to space index i . Note that the factor of 0.4π in equation (1) arises from the choice of c.g.s. units. Equation (1) is written in terms of flux density and magnetic field intensity. The relationship between these quantities is implicit and is supplied to the computer as data describing the magnetic characteristics of the material.

Assuming the boundary and initial conditions

$$B_z(r,0) = 0 \quad (2)$$

$$H_z(a,t) = \text{known driving function} \quad (3)$$

$$\left. \frac{\partial B_z(r,t)}{\partial r} \right|_{r=a} = 0 \quad (4)$$

*Partially supported by the National Science Foundation under Grants GP-3086 and GP-14509 and by the National Aeronautics and Space Administration under a Pre-Doctoral Fellowship.

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an iterative procedure is implemented to determine $B_z(r,t)$. It should be noted that in this procedure, B_z and H_z are related, not by a linear or other closed-form mathematical approximation, but by the data points describing the D.C. hysteresis loop of the material being considered.

Eddy Current Losses

When Ampere's Circuital Law is applied to the path 1,2,3,4,1 of Figure 1, equation (5) is obtained,

$$\frac{\partial H_z(r,t)}{\partial r} = -0.4 \pi J_0^2(r,t) \quad (5)$$

where $J_0(r,t)$ is the current density as a function of space and time. The time average squared current is then:

$$\overline{J_0^2(1)} = \frac{1}{T} \int_0^T J_0^2(r,t) dt \quad (6)$$

or, in difference terms,

$$\overline{J_0^2(1)} = \frac{1}{T} \sum_{j=1}^T J_0^2(j) \quad (7)$$

Next, the power density may be written as

$$W_E'(1) = \sigma J_0^2(1) \quad (8)$$

where W_E' is in units of watt cm^{-3} , if J is in amp cm^{-2} and σ is in ohm cm .

If the space indexing is so arranged that $i=2$ corresponds to $r=0$, then one observes that

$$r = (i-1.5) \Delta r \quad (9)$$

is the mean radius of the segment between r and $r+\Delta r$. Equation (8) may now be employed to determine the power per unit length of rod by multiplying equation (8) by an element of area in a z -constant plane. With reference to Figure 1, one will note that the desired element of area is

$$dA = \pi dr dy \quad (10)$$

In difference notation, when the azimuthal angle θ is allowed to vary over one revolution, equation (9) may be applied to equation (10) to yield

$$dA = 2\pi(i-1.5)(\Delta r)^2 \quad (11)$$

The power per unit length, W_E , is thus

$$W_E = 2\pi\sigma(\Delta r)^2 \overline{(i-1.5)J_0^2(i)} \quad (12)$$

HYSTERESIS LOSSES

The energy loss per hysteresis loop cycle is

$$\text{Energy} = \oint B \, dh \quad (13)$$

Power loss in watts, employing c.g.s. magnetic units, is

$$W_H' = \frac{f}{4\pi \times 10^7} \oint B \, dh, \quad (14)$$

where f is the frequency in cycles sec⁻¹, B is the flux density in gauss, H is the magnetic intensity in oersteds and W_H' is power loss in the watt cm⁻³. Casting equation (14) in difference form yields

$$W_H'(i) = \frac{f}{4\pi \times 10^7} \Delta H(i) \sum_{\lambda} \Delta B_{\lambda}(i) \quad (15)$$

where the sum-product, $\Delta H(i) \sum_{\lambda} B_{\lambda}(i)$, represents the area of the hysteresis loop numerically analyzed. If equation (15) is now multiplied by equation (11) the hysteresis loss at a given depth, i , per unit length of rod is obtained:

$$W_H'(i) = \frac{2\pi i(1-1.5)(nr)^2}{4\pi \times 10^7} \left[\Delta H(i) \sum_{\lambda} B_{\lambda}(i) \right]. \quad (16)$$

The total power loss due to hysteresis is the sum of terms of the type given in equation (16), with the surface ($i=19$) and the center ($i=2$) of the cylinder weighted one half to account for their reduced area:

$$W_H = \frac{2f}{4 \times 10^7} \left[\frac{1}{2} \sum_{i=2,19} (i-1.5) \Delta H(i) \sum_{\lambda} \Delta B_{\lambda}(i) + \sum_{i=3}^{18} (i-1.5) \Delta H(i) \sum_{\lambda} \Delta B_{\lambda}(i) \right]. \quad (17)$$

The evaluation of the sum, $\Delta H(i) \sum_{\lambda} B_{\lambda}(i)$, is interesting and worthy of elaboration: The hysteresis loops describing B - H variation at each space point in the material are stored in the memory of the digital computer as a result of the procedure discussed below. If one considers the maximum, $H_{\max}(i)$, and the minimum $H_{\min}(i)$, extremities of magnetization at each space point, $\Delta H(i)$ may be simply defined as

$$\Delta H(i) = H_{\max}(i) - H_{\min}(i) / 100, \quad (18)$$

where the factor 100 has been arbitrarily selected. It might be noted that on the major, saturated, hysteresis loop

$$\Delta H(i) = 2H_c / 100, \quad (19)$$

where H_c is known as the coercive field.

Since the procedure described below for the determination of the hysteresis loops yields only the left side of the hysteresis curve, which is assumed to be centrally

-4-

symmetric, it will be observed that the $\Delta B_\lambda(i)$ corresponding to each $\Delta H(i)$ may be evaluated as the sum of the $B_\lambda(i)$ corresponding to $-H(i)$. This is illustrated in Figure 2. Thus, the area of the hysteresis loop is evaluated as

$$\text{Area} = \Delta H(i) \sum \frac{\Delta B_\lambda(i)}{\lambda} \quad (20)$$

DETERMINATION OF MINOR HYSTERESIS LOOPS

The determination of minor hysteresis loops is necessary because the maximum flux density decreases with depth into the material and it is important to develop a procedure for determining on which hysteresis loop such a variation is occurring. The loop on which variation is taking place is a member of a family consisting of a very large number of minor loops.

The authors, therefore, have derived a relationship which is an extension of the one developed in references (2) and (4), which, when solved by a digital computer, makes it possible to obtain the data points describing any minor hysteresis loop from the given data of the dc major saturated hysteresis loop and the dc magnetization curve. This data provides the double valued relationship between flux density and magnetic intensity used implicitly in the finite difference equations.

In this development, it is assumed that the dc magnetization curve is the loci of the vertices of the nested symmetrical minor hysteresis loops. When an initially demagnetized sample is subjected to an applied magnetic field, the B-H relationship in the material is assumed to be described by the normal magnetization curve, curve A of Figure 3, until a local maximum is reached. When the amplitude of the applied magnetic field begins to decrease, the B-H relationship in the material is no longer described by the normal magnetization curve, but rather by one of the family of hysteresis loops. If the local maximum, point I of Figure 3, is equal to or greater than the saturation flux density, the hysteresis loop described will be the major saturated hysteresis loop; curve B of Figure 3. If the local maximum is less than the saturation flux density, point II of Figure 3, the hysteresis loop described will be one of the nested minor hysteresis loops. For cases where the local maximum, point II of Figure 3, is less than 0.01 of the saturation flux density the minor hysteresis loop will be represented by the normal magnetization curve; the following procedure, therefore, will be bypassed. This approximation is employed to speed up computer time without sacrificing accuracy.

The method by which any of these minor hysteresis loops is determined using known data describing the major saturated hysteresis loop and the dc magnetization curve is described below with reference to Figure 3.

In Figure 3, point II, on the normal magnetization curve (A), represents a local maximum less than the saturation flux density. The first step in the procedure is to locate point II* which is centrally symmetric with respect to point II. Having located point II* the next step is to scale the left curve of the major hysteresis loop horizontally so that it passes through point II*. This curve is represented by the dashed lines, curve D. The horizontally scaled major hysteresis curve is then scaled vertically with respect to point II* so that it passes through point II* and its centrally symmetric point, II. The resultant curve, curve E, represents the left half of the desired minor hysteresis loop.

CALCULATIONS AND TESTS

Actual computer calculations were carried out on the University of Pittsburgh's IBM 7090 digital computer. Quantities computed included flux density, eddy current density and hysteresis loss intensity at various depths in the rod, total eddy current loss, and total hysteresis loss. A number of reliable tests were needed to provide checks on the theoretical work. A 5/8 inch round rod of #1010 steel was shaped into a 11-7/16 inch average diameter ring, welded, and annealed to form a homogeneous continuous magnetic path. A primary or driving winding was wound on the ring and excited by a sinusoidal 60 c.p.s. source; tests were conducted with the device in an oil bath to make certain that the temperature of the specimen was known and remained nearly constant.

Test results are compared in Figure 4, not only with the calculations made by the procedures set forth in this paper, but also with Gillott's previous work¹ which neglected hysteresis phenomena. The reader will note that when hysteresis is considered that the eddy current portion of the losses decreases, but the total calculated losses increases to more closely match the experimentally determined total losses.

The calculated values of eddy current loss, hysteresis loss, and total core loss are given below for three values of applied magnetic field intensity.

$H_{max}(a,t)$ (oersteds)	P_e (watts)	P_h (watts)	P_T (watts)
20	9.60	5.32	14.92
50	49.14	6.78	55.92
60	68.34	7.10	75.44

SUMMARY AND CONCLUSIONS

The equation and results presented in this paper are for a cylindrical rod subjected to a sinusoidally varying 60 cycles per second magnetic field. The basic equations, however, were developed for solid ferromagnetic plates and cylinders², and the computer procedure is applicable to both periodic and non-periodic driving functions.

To the authors' knowledge, this represents a unique procedure for accurately calculating eddy current losses and hysteresis losses in ferromagnetic materials. Some of the engineering applications of this work are the losses in end plates, through bolts and finger plates of rotating machines; losses in solid iron rotors of turbine generators, some rotating induction machines, and eddy current brakes. Stationary electrical equipment for the heating of billets, electrical furnaces, and equipment for certain heat treating processes suggest further applications for the results of this present paper.

APPENDIX: COMPUTER PROGRAM

Language: MAD (Michigan Algorithmic Decoder)
Computer: IBM 7090
Monitor: University of Michigan Executive System

<u>Variable Name</u>	<u>Description</u>
B(:)	B(i,j)
BCK	B one cycle previous (used in STATE).
EDATA, HDATA	Data set describing left side of major saturated hysteresis loop.
BMAJ	Vertical transfation reference in DETL.
B PREV(I)	B(i,j-1)
BREF, HREF	Local maximum in DETL.
BTRANS, HTRANS	Scaled curve in DETL.
BVIRG, HVIRG	Data set describing virgin magnetization curve.
CLOCK ϕ	Starting time of program.
C ϕ N	$\frac{(\Delta t) \rho 10^8}{0.4\pi(\Delta r)^2}$
C ϕ NVG	Steady state factor
DELTA B(I)	$\Delta B(i)$
DELTA H(I)	$\Delta H(i)$
DELTA R	Δr
FLAG	Steady state determination index.
GL	Index in TERP ϕ .
CQ	Index in TERP ϕ .
H(I)	H(i,j)
HMAJ	Horizontal translation reference point in DETL.
I	i(space index)
J(I)	$J_e(r,t)$
J SQ(I)	$J_e^2(r,t)$
L	Ordering index on XCAL, YCAL.
L ϕ X, HI X	Limits on integration of hysteresis loop.
M	Count index on XCAL, YCAL.
PWR EDY(I)	$W_E'(i)$
PWR HYS(I)	$W_H'(i)$

<u>Variable Name</u>	<u>Description</u>
Q	Index in TERP0.
QQ	Ordering index in TERP0.
RH0	$\rho \times 10^8$
SUM DEL(I)	$\sum B_A(i)$
T	t
TOT EDY	W_E
TOT HYS	W_H
TOT PWR	W_T
XCAL, YCAL	List of B-H points currently in use at a space point.
XIN(T)	$H(a,j)$
X POINT	H value used in integrating hysteresis loop.
XUSE, YUSE	List used in TERP0.
Z	Final cycle index.

<u>Function</u>	<u>Description</u>
.ABS. X	$ X $
.AND.	(Logical).
CLOCK.	Reads the time clock attached to the computer.
COS.(x)	cosine(x)
DETL.	Procedure to determine on which hysteresis loop variation is occurring.
DOUBLE.	Loading procedure.
.E.	=
F TRAP.	Converts an underflow to zero.
ERROR.	Initiates error sequence in monitor.
.G.	>
.L.	<
.OR.	(Logical) +
OUTPUT.	Determines W_E , W_H and W_T
STATE.	Tests for steady state by comparing B values one cycle apart.
SYSTEM.	Returns control to monitor.
TERP0.	Linear interpolation function.

Remarks:

<u>Statement number(s)</u>	<u>Comments</u>
001,089	Indicates running time of program
012-017	Check if convergence criteria is met
022-024	Loading of $H(a,t)$
025-028	Initializing all space points to virgin magnetization curve
029,030	Begins procedure
035,042	Tests if TRP has modified XCAL, YCAL
038,050	Tests proper B-H relationship
061	Prints output at steady state check points
062	Prints complete output during last cycle
101	$J_e^c(i)$ is implicitly determined
101	$\rho = RH\theta \times 10^{-8}$
106-110	Evaluation of $\Delta H(i)/AB_1(i)$
155-163	Restoration of virgin magnetization curve
215	Zero width hysteresis loop when local maximum is small

\$ ZERO, COMPILE MAD, PUNCH OBJECT, EXECUTE, CUMP

MAD (24 SEP 1964 VERSION) PROGRAM LISTING

```
CLOCK O =CLOCK.(0)          *001
PRINT COMMENT $1 TOTAL LOSSES IN CYLINDRICAL RODS $      *002
F TRAP.                  *003
INTEGER K,I,T,N,M,Z,II,CLOCK,-CLOCK C,FLAG            *004
DIMENSION M(44),L(19),BPREV(19),H(19),J(19),
1    YCAL(19*60),XCAL(19*60),BDATA(44),HDATA(44) ,      *005
2    BVIRG(30),HVIRG(30),XIN(360),B(19),B CK(4),      *005
3    J SQ(19),PWR EDY(19),DELTA H(19),SUM DEL(19),PWR HYS(19)
READ AND PRINT DATA          *006
READ FORMAT INPUT,(K=2,1,K.G.30,BVIRG(K),HVIRG(K))      *007
READ FORMAT INPUT,(K=2,1,K.G.44,BDATA(K),HDATA(K))      *008
VECTUR VALUES INPUT=S 2F12.4* S          *009
DELTA R= RADIUS/17.          *010
PI=3.1415926                *011
WHENEVER (RHO*(1.+COS.(PI/(36.*DELTA R))))>G.        *012
4    (0.4*PI*360.*F*DELTA R.P.2*BVIRG(2)/HVIRG(2))      *012
    PRINT COMMENT $6 NUMERICAL SOLUTION UNSTABLE, EXECUTION
1 TERMINATED $              *013
    SYSTEM.                  *014
OTHERWISE                   *015
    PRINT COMMENT $6 NUMERICAL SOLUTION ASSUMED STABLE $      *016
END OF CONDITIONAL          *017
CON= RHO/1F*360.*0.4*PI*DELTA R .P.2
PRINT RESULTS CON          *018
DEL PHI= PI/180.             *019
THROUGH INX, FOR T=0,1,T.G.360      *020
XIN(T)=PEAK*SIN.(PHI)          *021
PHI= PHI+ DEL PHI          *022
INX                         *023
    THROUGH VIRGIN,FOR I=2,1,I.G.19      *024
DOUBLE.(I)
L(I)=-1                      *025
VIRGIN                      *026
                            *027
                            *028

R WHEN L=-1, K=2 OCCURS AT THE UPPER RIGHT END OF
R THE HYSTERESIS CURVE. WHEN L=+, K=2 OCCURS AT
R THE LOWER LEFT END.

I=18                      *029
T=C                      *030
LCOP                      *031
WHENEVER I.E.18          *032
    H(I+1)=XIN(T)          *033
FIND M2      M (I)=TERPO.(B (I),YCAL,XCAL,2,P(I),3,1)      *034
                WHENEVER H(I).G.1E2%, TRANSFER TO FIND M2      *035
                B PREV(I+1)=B(I+1)          *036
                B(I+1)=TERPO.(XIN(T+1),XCAL,YCAL,2,M(I+1),2,I+1)      *037
                DETL.(I+1)          *038
END OF CONDITIONAL        *039
```

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```

WHENEVER I.G.2                                •040
FIND H1      'C(I-1)=TERPO.(B(I-1),YCAL,XCAL,2,M(I-1),3,I-1)   •041
              WHENEVER H(I-1) .G. 1E29, TRANSFER TO FIND H1   •042
END OF CONDITIONAL                            •043
BPREV(I)=B(I)                                 •044
WHENEVER I.G.2                                •045
      B(I)=((CON*(H(I+1)*(1.+(1./(2.*(I-2)))))-2.*H(I))  •046
      +H(I-1)*(1.-(1./(2.*(I-2)))))+B(I)                  •046
  OTHERWISE                                     •047
      B(I)=(2.*CON*(H(I+1)-H(I)))+B(I)                   •048
END OF CONDITIONAL                            •049
DETL.(I)                                      •050
WHENEVER FLAG.E.1                             •051
      J(I)=(H(I)-H(I+1))/(DELTA R + 0.4 * PI)           •052
      WHENEVER Z.E.0, J SQ(I)=J SQ(I)+J(I).P.Z          •053
      WHENEVER Z.E.1 .AND. T.E.C, EXECUTE OUTPUT.       •054
END OF CONDITIONAL                            •055
WHENEVER T.GE.3                                •056
      WHENEVER N.G.11 .AND. FLAG.NE.1 .AND. T.E.15, EXECUTE STATE.  •057
      T=T+1                                         •058
      TRANSFER TO LCOP                           •059
END OF CONDITIONAL                            •060
WHENEVER (T.E.0 .OR. T.E.90 .OR. T.E.180 .OR. T.E.270) .AND.  •061
  FLAG.NE.1, PRINT RESULTS N,T,B(2)...B(19)
WHENEVER FLAG.E.1, PRINT RESULTS N,T,B(2)...B(19)          •062
WHENEVER T.L.359                               •063
      T=T+1                                         •064
  OTHERWISE                                     •065
      T=0                                           •066
      N=N+1                                         •067
      WHENEVER Z.E.0,Z=1                          •068
      WHENEVER FLAG.E.13,Z=0                      •069
      WHENEVER FLAG.E.13,FLAG=1                  •070
END OF CONDITIONAL                            •071
  T=T+1                                         •072
  TRANSFER TO LOOP                           •073

```

R INTERNAL FUNCTION "STATE." DETERMINED WHEN STEADY
 R STATE CONDITIONS HAVE BEEN ACHIEVED.

INTERNAL FUNCTION	•C74
ENTRY TO STATE.	•075
WHENEVER T.E.0	•076
WHENEVER .ABS.(B CK(1)-B(I)).L..ABS.(CONVG*B(I)), FLAG=10	•077
B CK(1)=B(I)	•078
OR WHENEVER T.E.90	•079
WHENEVER .ABS.(B CK(2)-B(I)).L..ABS.(CONVG*B(I)) .AND.	•080
FLAG.E.10, FLAG =11	•081
B CK(2)=B(I)	•082
OR WHENEVER T.E.180	•083
WHENEVER .ABS.(B CK(3)-B(I)).L..ABS.(CONVG*B(I)) .AND.	•084
FLAG.E.11, FLAG=12	
B CK(3)=B(I)	

```

    .01 WHENEVER T.E.270          *085
      WHENEVER .ABS.(B CK(4)-B(1)).L..ABS.(CONVG+B(1)) .AND.
1       FLAG.E.12               *086
        FLAG=13                 *087
        PRINT COMMENT $6 STEADY STATE ACHIEVED AT$ *088
        MINUTE=CLOCK.(C)-CLOCK D                *089
        PRINT RESULTS MINUTE, N, 1              *090
        PRINT COMMENT $6$                      *091
        END OF CONDITIONAL                   *092
        B CK(4)=B(1)                      *093
        END OF CONDITIONAL                   *094
        FUNCTION RETURN                     *095
        END OF FUNCTION                    *096

R     INTERNAL FUNCTION 'OUTPUT.' DETERMINES AND PRINTS
R     HYSTERESIS AND EDDY CURRENT LOSSES.

INTERNAL FUNCTION                         *097
INTEGER LO X, HI X                      *098
ENTRY TO OUTPUT.                         *099
THROUGH POWER, FOR I=2,1,I.G.18         *100
PWR EDY(I)=RHO*(1.E-8)*(J SQ(I)/360.)*
1     (DELTA R.P.2)*2.*PI*(I-1.5)       *101
POWER
LO X=6                                 *102
HI X= M(I)-5                          *103
THROUGH LOSS, FOR I=2,1,I.G.19         *104
DELTA H(I)= .ABS.(XCAL(I,LO X) - XCAL(I,HI X))/100. *105
THROUGH HYSTRS, FOR XPOINT= .ABS.(XCAL(I,LO X)), -DELTA H(I),
1     XPOINT.E. -1.*.ABS.(XCAL(I,HI X))      *106
1     DELTA B=.ABS.(TERPO.(XPOINT,XCAL,YCAL,2,M(I),2,I) +
1             TERPO.(-XPOINT,XCAL,YCAL,2,M(I),2,I))   *107
1     SUM DEL(I)= SUM DEL(I)+ DELTA B        *108
HYSTRS
PWR HYS(I)=DELTA H(I)+SUM DEL(I)*F*(I-1.5)*
1     (DELTA R.P.2)/(2.E7)                  *109
LOSS
THROUGH TOTAL 1, FOR I=2,1,I.G.18      *110
TOT EDY=TOT EDY + PWR EDY(I)           *111
TOTAL 1
THROUGH TOTAL 2, FOR I=3,1,I.G.18      *112
TOT HYS= TOT HYS+ PWR HYS(I)           *113
TOTAL 2
TOT HYS= TOT HYS+0.5*(PWR HYS(2)+PWR HYS(19)) *114
TOT PWR = TOT HYS+ TOT EDY             *115
PRINT COMMENT $2 FINAL RESULTS $      *116
PRINT RESULTS PWR EDY(2)...PWR EDY(18), PWR HYS(2)...PWR HYS(19),
1     TOT EDY, TOT HYS, TOT PWR          *117
SYSTEM.
FUNCTION RETURN                       *118
END OF FUNCTION                      *119

```

R INTERNAL FUNCTION 'TERPO.' PERFORMS A LINEAR
R INTERPOLATION ON A LIST.

R X IS THE INDEPENDENT VARIABLE
R XT IS THE NAME OF THE LIST OF INDEPENDENT VARIABLES
R YT IS THE NAME OF THE LIST OF DEPENDENT VARIABLES
R LOLIM IS THE LOWER LIMIT ON THE INDEX
R HILIM IS THE UPPER LIMIT ON THE INDEX
R ORDER IS THE INDEX ON LOADING
R FIXED IS THE VALUE OF THE FIXED SUBSCRIPT
R Y IS THE INTERPOLATED DEPENDENT VARIABLE

R TERPO. - VERSION 42

INTERNAL FUNCTION (X,XT,YT,LOLIM,HILIM,ORDER,FIXED)
DIMENSION XUSE(70),YUSE(70) *126
INTEGER LOLIM,HILIM,ORDER,FIXED,G,QQ,GL,GQ *127
ENTRY TO TERPO. *128

WHENEVER ORDER.E.G
THROUGH IN1, FOR Q=LOLIM,1,Q.G.HILIM
XUSE(Q)=XT(Q) *129
YUSE(Q)=YT(Q) *130

IN 1

OR WHENEVER ORDER.E.2 .OR. ORDER.E.3
THROUGH IN2, FOR Q=LOLIM,1,Q.G.HILIM *131
XUSE(Q)=XT(FIXED,0) *132
YUSE(Q)=YT(FIXED,Q) *133

IN 2

OTHERWISE
PRINT COMMENT\$6*** ERROR RETURN FROM ORDER\$
ERROR. *141

END OF CONDITIONAL *142

Q=(LOLIM+HILIM)/2 *143
QQ=(XUSE(HILIM)-XUSE(LOLIM))/ABS.(XUSE(HILIM)-XUSE(LOLIM)) *144
WHENEVER XUSE(Q).L.X *145

GL=-1 *146

OR WHENEVER XUSE(Q).G.X *147

GL=1 *148

OTHERWISE *149

FUNCTION RETURN YUSE(Q) *150

END OF CONDITIONAL *151

T.S.

WHENEVER Q.G.HILIM .OR. Q.L.LOLIM *152

WHENEVER ORDER .E.3 .AND. |(FIXED)|.NE.58 *153

PRINT COMMENT SC LOLIM OR HILIM HAS BEEN EXCEEDED. *154

INTERPOLATION WILL CONTINUE ON THE VIRGIN CURVE. S *155

PRINT RESULTS XUSE(LOLIM), XUSE(HILIM), X,I,N,T *156

DOUBLE.(FIXED) *157

OTHERWISE *158

PRINT COMMENT\$** ERROR RETURN FROM QS *159

TRANSFER TO OUT *160

END OF CONDITIONAL *161

FUNCTION RETURN 22E33 *162

*126
*127
*128
*129
*130
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OR WHENEVER .ABS. QQ.NE.1 *164
PRINT COMMENTS\$6*** ERROR RETURN FROM QQS *165
OUT PRINT RESULTS LOLIM,HILIM,XUSE(LCLIM),XUSE(HILIM),Q,XUSE(Q), *166
1 X,GL,ORDER,QQ,FIXED,M(FIXED) *166
ERROR. *167
END OF CONDITIONAL *168
WHENEVER XUSE(Q).L.X *169
WHENEVER GL.L.Q *170
Q=Q+QQ *171
TRANSFER TO TEST *172
OTHERWISE *173
TRANSFER TO COMP *174
END OF CONDITIONAL *175
OR WHENEVER XUSE(Q).G.X *176
WHENEVER GL.G.Q *177
Q=Q-QQ *178
TRANSFER TO TEST *179
OTHERWISE *180
TRANSFER TO COMP *181
END OF CONDITIONAL *182
OTHERWISE *183
FUNCTION RETURN YUSE(Q) *184
END OF CONDITIONAL *185
COMP GU=Q+QQ*GL *186
FUNCTION RETURN YUSE(Q)+((X-XUSE(Q)) \times (YUSE(GQ)-YUSE(Q))/ *188
1 (XUSE(GQ)-XUSE(Q))) *188
END OF FUNCTION *189

R INTERNAL FUNCTION 'DOUBLE.' LOADS THE VIRGIN CURVE *190
R INTO THE XCAL,YCAL ARRAY IN BOTH FIRST AND THIRD *191
R QUADRANTS. *192

INTERNAL FUNCTION (FIX) *193
INTEGER FIX,V *194
ENTRY TO DOUBLE. *195
K=2 *196
THROUGH VRG 1, FOR V=2,1,V.G.58 *197
WHENEVER V.L.30 *198
XCAL(FIX ,V)=-1*HVIRG(K) *199
YCAL(FIX ,V)=-1*BVRG(K) *200
K=K+1 *201
OR WHENEVER V.GE.30 *202
XCAL(FIX ,V)=HVIRG(K) *203
YCAL(FIX ,V)=BVRG(K) *204
K=K-1 *205
END OF CONDITIONAL *206
VRG 1 M(FIX)=58 *207
FUNCTION RETURN
END OF FUNCTION

R INTERNAL FUNCTION DETL.

R DETL. COMPARES THE CURRENT VALUES OF B AND H WITH
 R HSAT AND HSAT. IT RETURNS THE CORRECT MINOR LOOP
 R IN XCAL,YCAL. THIS MINOR LOOP WILL HAVE B ,H AS ITS
 R VERTEX, AND THE VERTEX WILL LIE ON THE VIRGIN CURVE.

R DETL. VERSION 4C

INTERNAL FUNCTION (INDEX)	*208
INTEGER K,L,M,INDEX	*209
DIMENSION HTRANS(44),BTRANS(44)	*210
ENTRY TO DETL.	*211
WHENEVER L<INDEX)*B<INDEX).LE. L(INDEX)*B PREV(INDEX), L(INDEX)=L(INDEX) FUNCTION RETURN *	*213
BREF=.ABS.B (INDEX)	*214
WHENEVER BREF.L=(BSAT/100.), FUNCTION RETURN	*215
HREF=.ABS.H (INDEX)	*216
L(INDEX)=44	*217
WHENEVER BREF.GE.BSAT	*218
THROUGH BACK 1, FOR K=2,1,K.G.44	*219
YCAL(INDEX,K)=BDATA(K)*L(INDEX)	*220
XCAL(INDEX,K)=HDATA(K)*L(INDEX)	*221
 BACK 1	
FUNCTION RETURN	*222
END OF CONDITIONAL	*223
HMAJ=TERPO.(-BREF,BDATA,HDATA,2,44,0,0)	*225
HFAC=2*HREF/(HREF+.ABS.HMAJ)	*226
THROUGH TRAN 1, FOR K=2,1,K.G.44	*227
HTRANS(K)=(HDATA(K)+HREF)*HFAC-HREF	*228
 TRAN 1	
3MAJ= TERPO.(HREF,HTRANS,BDATA,2,44,0,0)	*229
BFAC=2*BREF/(BREF+.ABS.BMAJ)	*230
THROUGH TRAN 2, FOR K=2,1,K.G.44	*231
BTRANS(K)=(BDATA(K)+BREF)*BFAC-BREF	*232
 TRAN 2	
THROUGH BACK2, FOR K=2,1,K.G.44	*233
XCAL(INDEX,K)=HTRANS(K)*L(INDEX)	*235
YCAL(INDEX,K)=BTRANS(K)*L(INDEX)	*236
 BACK 2	
FUNCTION RETURN	*237
END OF FUNCTION	*238
END OF PROGRAM	*239
	*240
	*241

ACKNOWLEDGMENTS

The authors wish to express their thanks to the United States Steel Research for the preparation of the magnetic sample and express particular appreciation to Mr. James Hale and his colleagues at Allegheny Ludlum Research for making all the very carefully conducted loss tests, the results of which are shown in this paper.

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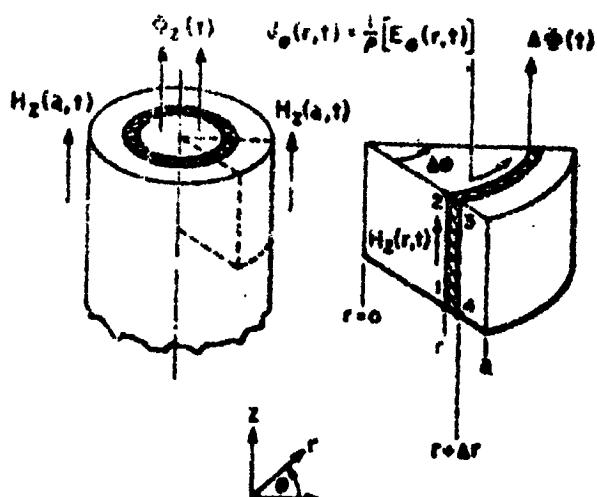


Figure 1. Geometry and field relationships of infinite cylindrical rod

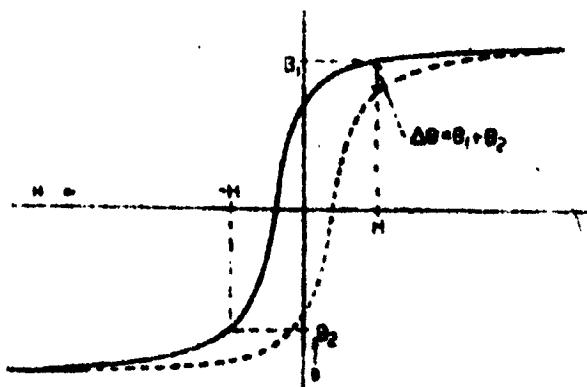


Figure 2. Numerical evaluation of the area enclosed by a hysteresis loop

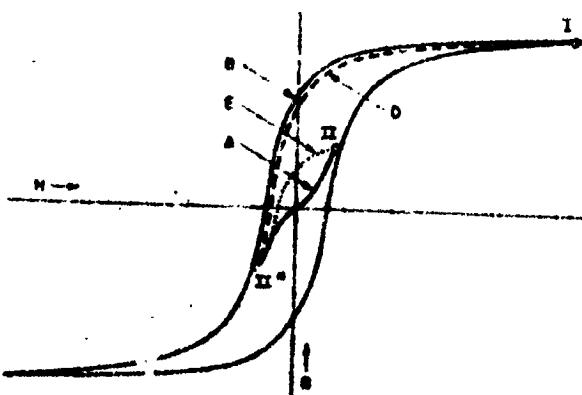


Figure 3. Determination of a minor hysteresis loop

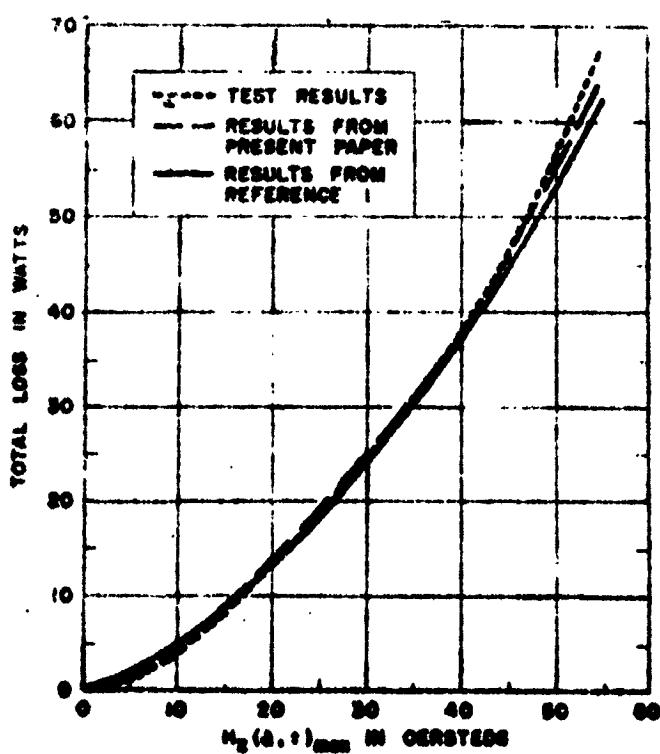


Figure 4. Comparison of calculated and experimental losses